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Patentanmeldung Nr.

Patent application No. Demande de brevet n°

03425218.9

Der Präsident des Europäischen Patentamts; -- Im Auftrag

For the President of the European Patent Office

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Bezeichnung der Erfindung/Title of the invention/Titre de l'invention: (Falls die Bezeichnung der Erfindung nicht angegeben ist, siehe Beschreibung. If no title is shown please refer to the description.

Si aucun titre n'est indiqué se referer à la description.)

Method of generating a switching sequence for an unary array of conducting branches and a relative thermometrically decoded digital-to-analog conveter

In Anspruch genommene Prioriät(en) / Priority(ies) claimed /Priorité(s) revendiquée(s)
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FIELD OF THE INVENTION

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The present invention relates to digital-analog converters and in particular to a method of generating a switching sequence for an unary array of conducting branches and a relative thermometrically decoded digital-to-analog converter having a desired integral non-linearity (briefly INL) error function.

BACKGROUND OF THE INVENTION

Digital-analog (D/A) converters are widely used in many applications, such as wireless communications, signal reconstruction, waveforms generation etc. Basically, a D/A converter may be realized with an array of conducting branches that generate binary weighted constant currents (binary array) and/or identical currents (unary array). These conducting branches may include capacitors or current generators, selectable by respective switches for delivering current toward a summing line or toward ground.

For sake of simplicity, let us refer to a D/A converter, the conducting branches of which are realized with current generators, though what will be said will hold, mutatis mutandis, also for conducting branches realized with switched capacitors.

The conducting branches of a binary array are selected by respective bits of an input bit string, while the conducting branches of an unary array are selected according to a fixed switching sequence for making the unary array generate a total current corresponding to the input digital value.

For better comprehending the field of utility of this invention, let us refer to the so-called segmented converters, which generally have a binary weighted array of conducting branches, selected by the least significant bits of an input digital string, and an unary array that delivers a current corresponding to the digital value represented by the most significant bits, that is according to a so-called thermometer or thermometric decoding.

A sample unary array is depicted in Figure 1. As may be noticed, the current generators are selected according to a certain switching sequence and the output

current Iout corresponds to the sum of the currents circulating in the selected branches. Optionally, there may be a "dummy" area in the unary array, which may be used for realizing biasing circuits. In segmented digital-to-analog converters, the "dummy" area may be used for realizing a binary array.

When the digital input value to be converted by the unary array represents a 5 certain number n, the current generators of the unary array from 1 to n are switched on, thus generating an output current lout proportional to the digital input number n.

Ideally, a D/A converter should generate an output signal that varies linearly with the input bit string, which would happen if all the conducting branches were identical. Unfortunately, mismatches between conducting branches due to inaccuracies of the fabrication process (process spread), make the current delivered by each branch not exactly equal to the design value, but affected by an error that may depend on the position of the conducting branch on the silicon substrate. 15

In general, unary arrays are affected by a differential non-linearity (DNL) error and by an integral non-linearity (INL) error.

Indicating with \overline{I} the average current delivered by the branches of an unary array and with I_j the current delivered by the j-th branch of the array in the switching sequence,

$$I_{j} = \overline{I} \cdot \left(1 + \varepsilon_{j}\right)$$

wherein ε_j is the relative deviation of the current I_j from the average current \overline{I} .

In an unary array not having a dummy area, the DNL error of the k-th branch in the switching sequence is

$$DNL(k) = \varepsilon_k$$

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which represents a non-uniform deviation in the ideal current step amplitude between adjacent bit strings.

The INL error function is defined as

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$$INL(k) = \sum_{j=1}^{k} \varepsilon_{j}$$

and gives the deviation of the real analog output signal from its ideal value as a percentage of the average current \overline{I} for any value "k" of the switching sequence.

The absolute INL error of a switching sequence is the maximum absolute value of the relative INL error function.

An introduction on the INL and DNL errors of an unary array of a D/A thermometrically decoded converter is carried out in the article by Y. Cong and R. L. Geiger "Switching Sequence Optimization for Gradient Error Compensation in Thermometer-Decoded DAC Arrays", *IEEE Trans. on circuits and systems - II:* analog and digital signal processing, Vol. 47, No. 7, pages 585-595, July 2000.

The DNL error of a branch can be reduced only by reducing the process spread and is independent from the switching sequence. By contrast, the INL error function strongly depends on the switching sequence, as it may be easily inferred from the following example.

Two possible switching sequences of a mono-dimensional unary array having eight conducting branches are shown in Table 1. The conducting branch are affected by the indicated DNL errors (ε). The absolute values of the underlined numbers are the absolute INL errors of the switching sequences.

As may be noticed, the INL error function of the sequential switching sequence shows a maximum deviation of the analog output from its ideal value in the middle of the sequence. This situation is inconvenient, because the digital values input to a D/A converter are more likely in the middle of the range of conversion, rather than at the two ends thereof. Therefore, it is more convenient to switch the

mono-dimensional array of conducting branches of Table 1 according to a symmetrical sequence than according to a sequential sequence. Moreover, the absolute INL error for the symmetrical sequence is 7, while for the sequential sequence is 16.

- In this very simple case, it is possible to determine by successive trials the switching sequence with the smallest absolute INL error, but for two-dimensional unary arrays of about a thousand of conducting branches the number of combinations is too large for determining a switching sequence with the desired INL error function by trials.
- 10 Many different methods of determining a switching sequence of a twodimensional unary array of conducting branches and a relative D/A converter have been proposed.

- The patent US 6,118,398 by G.J. Fisher et al. discloses a digital-analog converter having an unary array of current sources that are selected according to a sequence that ensures a relatively small absolute integral non-linearity error. The suggested switching sequence is substantially a mixed symmetrical sequence, in which the current sources that are in the middle of the array have median positions in the switching sequence, while current sources that are in borderline regions of the array are at the beginning or at the end of the sequence.
- The patent US 5,057,838 by K. Tsuji et al. discloses a D/A converter having a plurality of conducting branches of a two-dimensional array, wherein the switching sequence is determined in order to make the center of the current contributions delivered by the conducting branches of the array coincide with the center of the array.
- The document "A 14-bit Intrinsic Accuracy Q2 Random Walk CMOS DAC" by Van der Plas, Steyaert et al., JSCC 12 Dec '99, discloses a method of determining the switching sequence of the switches of a D/A converter organized in matrix form, exploiting the so-called "Q² random walk" algorithm.

The document "Switching Sequence Optimization for Gradient Error Compensation in Thermometer-Decoded DAC Arrays", by Cong, Geiger, JSSC 7 Jul '00, discloses an algorithm to find the so-called "INL bounded" switching sequence for an unary array of branches organized in matrix form.

5 Unfortunately, the absolute INL errors of the D/A converters realized with the above techniques remain relatively large.

OBJECT AND SUMMARY OF THE INVENTION

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As it will be more clearly described hereinafter, the array of conducting branches is affected by an error whose spatial distribution has an anti-symmetrical linear component, a quadratic symmetrical component, components of higher orders and a random component. The known thermometrically decoded converters are affected by quite large INL errors because the switching sequence of their unary array of conducting branches is not determined in a way that would optimally compensate the quadratic component of the error distribution.

15 The objective of the method of this invention is to determine switching sequences of two-dimensional unary arrays of conducting branches of thermometrically decoded D/A converters, in a way that will ensure that the relative INL error function be contained between pre-established symmetrical upper and lower bound functions. These functions may be constant, such to ensure an absolute INL error smaller than a pre-established value, or approaching zero in correspondence of mid way values of the switching sequence from a certain maximum value in correspondence of the two ends of the range of conversion. This last solution is preferable when the D/A converter is likely to work almost constantly about the middle of its range of conversion.

When these upper and lower bound functions are constant, the obtained switching sequence compensates both the linear and the quadratic component of the error distribution and therefore is affected by a very small absolute INL error, which depends essentially on the random component of the error distribution.

The method of this invention may be easily implemented by a computer program and allows the realization of thermometrically decoded D/A converters affected by a known limited INL error function.

The invention is more precisely defined in the annexed claims.

5 BRIEF DESCRIPTION OF THE DRAWINGS

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The advantages of this invention will become more evident through a detailed description referring to the attached drawings, wherein:

Figure 1 illustrates a sample 16x16 unary array;

Table 1 compares the INL error functions of two switching sequences for a mono-dimensional array;

Figure 2 shows a sample joint error distribution;

Figure 3 shows how to build a switching sequence according to the method of this invention;

Figure 4 shows a flow chart of a preferred embodiment of the method of this invention;

Figure 5 depicts a switching sequence determined by using the method of this invention for a 16x16 unary array;

Figure 6 depicts a sequential switching sequence for a 16x16 unary array;

Figure 7 depicts a symmetrical switching sequence for a 16x16 unary array;

Figure 8 depicts a switching sequence for a 16x16 unary array of the 12-bit thermometrically decoded D/A converter embedded in the commercial devices MTC-xx154, MTC-xx174 and MTC-xx454 of Alcatel Microelectronics;

Figure 9 depicts a switching sequence for a 16x16 unary array determined with a O^2 random walk algorithm;

Figure 10 depicts an "INL bounded" switching sequence for a 16x16 unary array;

Figure 11 depicts a random switching sequence for a 16x16 unary array;

Figure 12 depicts an anti-symmetrical switching sequence in respect to the center of a 16x16 unary array;

Figure 13 depicts an improved anti-symmetrical switching sequence in respect to

the center of a 16x16 unary array;

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Figures 14 and 15 compare the INL error functions of the switching sequences of Figures from 5 to 13 in function of the direction θ of the linear error distribution;

Tables 2 and 3 compare the maximum values of the absolute INL errors of many switching sequences obtained with known methods and the method of the invention for different values of the linear g_l and quadratic g_q coefficients;

Figure 16 depicts the switching sequence of an unary array of 1024 conducting branches of a 14-bit D/A converter of this invention, ten of which are thermometrically decoded.

10 DESCRIPTION OF AN EMBODIMENT OF THE INVENTION

Before illustrating the method of this invention, it is necessary to make some mathematical considerations.

As stated hereinbefore, the conducting branches of an array are affected by an error, which in electronic equipments may be due to thermal and/or mechanical phenomena, doping differences, and other process spread mechanisms of inaccuracy, distributed according to an error distribution function over the array.

In general, for a two-dimensional disposition of conducting branches, the relative error distribution $\varepsilon(x, y)$ may be approximated by a Taylor series expansion:

$$\varepsilon(x,y) = a_0 + a_{11}x + a_{12}y + a_{21}x^2 + a_{22}y^2 + a_{23}xy + \cdots$$

Truncating this series at the second order, neglecting the term $a_{23}xy$ and supposing that $a_{21}=a_{22}$, the linear $\varepsilon_l(x,y)$, quadratic $\varepsilon_q(x,y)$ and joint $\varepsilon_j(x,y)$ error distribution functions are, respectively,

$$\varepsilon_l(x,y) = g_l \cdot (\cos\theta \cdot x + \sin\theta \cdot y);$$
 $\varepsilon_q(x,y) = g_q \cdot (x^2 + y^2) - a_0;$ $\varepsilon_l(x,y) = \varepsilon_l(x,y) + \varepsilon_q(x,y)$

wherein θ is the angle of the linear error gradient, g_l is the slope, g_q is a quadratic

coefficient and a_0 is an offset value. This joint error distribution, represented in Figure 2, has proven to be a good approximation of real error distribution functions of unary arrays. The difference between the real error distribution and the joint error distribution is due to the components of higher order thereof and to a random error component.

In order to compensate errors due to the linear component of the error distribution, according to the method of the invention each pair of consecutive odd (2n-1) and even (2n) component of the switching sequence must be symmetrical in respect to the "center of gravity" of the two-dimensional array. In fact

$$\varepsilon_l(x,y) = -\varepsilon_l(-x,-y)$$

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and thus, if pairs of consecutive branches $\{a,b\}$ and $\{g,h\}$ are disposed as depicted in Figure 3, the linear component of the error introduced by each pair is null.

Being $\overline{\varepsilon_q}$ the continuous component of the quadratic error distribution over the area of the array of branches, that is the mean value thereof, the quadratic error distribution is also given by

$$\varepsilon_q(x,y) = \overline{\varepsilon_q} + \varepsilon_q^{AC}(x,y)$$

wherein $\varepsilon_q^{AC}(x, y)$ is the alternate component of the quadratic error.

The offset term $\overline{\varepsilon_q}$ does not contribute to the INL error because it causes only an offset error in the slope of the transfer function of the D/A converter. On the contrary, the alternate component has positive and negative values that make the transfer function nonlinear, thus originating the INL error.

Therefore it is clear that, in order to have a switching sequence with a desired (small) INL error function, it is necessary to compensate the alternate component of the quadratic error distribution.

According to an essential embodiment of the method of the invention, the first step consists in defining an upper bound function and a lower bound function, symmetrical to each other, of the INL error function of the switching sequence to be generated. Then the error distribution function over the array is evaluated, in order to calculate the error associated to each pair of symmetrical branches. As stated before, each pair of successive branches must be symmetrical in respect to the center of gravity in order to compensate the component of the INL due to the linear error distribution.

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These bound functions define the range of variation of the INL error and they may

be constant, if the absolute INL error must be minimized.

The appropriate switching sequence is built by choosing a first pair of branches (1, 2) starting from any branch of the array and performing iteratively the following steps:

- calculating a corresponding value of the INL error function of the switching sequence being built,
- choosing as the successive pair, the pair of conducting branches that maximizes or minimizes the next value of the INL error function of the switching sequence though remaining comprised between the corresponding values of the upper bound and lower bound functions,
- if all other pairs do not meet the preceding conditions, then repeating the steps restarting from the first by choosing every time a different first pair of branches, and if the conditions cannot yet be met, changing at least one of the bound functions and restarting from the first step.

Finally, when the determination of the appropriate switching sequence has been completed, it is output.

Preferably, the above described method is repeated for all possible first pairs of conducting branches, thus generating a set of switching sequences affected by an INL error function comprised between the same bound functions. The optimal switching sequence is chosen from this set according to a pre-established

criterion.

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Optionally, the bound functions may be closer to zero in correspondence of mid way values of the range of the converter compared to their value in correspondence of the two ends of the range of the converter.

A preferred embodiment of the method of this invention is described by the flow chart of Figure 4. In this case the symmetrical bound functions are constant and this constant value is half the maximum absolute value of the DNL error of the array.

An example of a 16x16 matrix describing a switching sequence determined according to the method of the invention illustrated in Figure 4 and for $g_l = g_q = 0.5$ is depicted in Figure 5. As may be noticed, the switching sequence has been determined by selecting every even conducting branch $(2, 4, 6 \dots)$ symmetrical to the preceding odd branch $(1, 3, 5 \dots)$ in respect to the center of the array.

Preferably, but not necessarily, the map of the switching sequence will be antisymmetrical in respect to an axis of symmetry of the array, that means that the last branch of the sequence is symmetrical to the first branch in respect to that axis of symmetry, and so forth for the other branches. For example, the positions of the branches 256, 255, 254 etc. of the switching sequence of Figure 5 are symmetrical to the positions of the branches 1, 2, 3 etc. in respect to the horizontal axis of symmetry of the array.

For comparison purposes, eight other matrices obtained with different algorithms are depicted in Figures from 6 to 13.

The resulting absolute INL error values in function of the angle θ of the gradient of the linear error are compared in Figures 14 and 15 for $g_l = g_q = 0.5$. The values have been calculated after having normalized to 1 the maximum value of the linear or quadratic errors. As may be noticed, the absolute INL error of the switching sequence of Figure 5 is always smaller than that of the other sequences.

Moreover, differently from the switching sequences of Figures from 6 to 13 obtained by using the known methods, the absolute INL error of the switching sequence of Figure 5 is substantially independent from the angle θ . This is very important because it ensures that the method of the invention is not tied to a particular shape of the error distribution, as the method disclosed in the aforementioned paper by Y. Cong and R. L. Geiger.

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Table 2 compares many switching sequence, indicating the maximum value of the absolute INL error for each of them. The references [1] and [2] indicate that the switching sequence is obtained using the method described in the mentioned paper by G. A. Van der Plas et al. and by Y. Cong and R. L. Geiger, respectively.

As may be noticed, the switching sequence of Figure 5 is the best one. This result is confirmed even using different values of g_l and g_q , as shown in Table 3. Even if these parameters undergo sensible variations, the maximum absolute INL error of the switching sequence obtained with the present method is substantially independent from them. This extraordinary result confirms that the method of this invention effectively compensates both the linear and the quadratic component of the error distribution.

The method of this invention may be applied whichever the shape of the array of conducting branches is, not only to two-dimensional square arrays.

A switching sequence generated according to the preferred embodiment of the method illustrated in Figure 4 for a substantially oval unary array of conducting branches of a 14-bit D/A converter of the invention, ten of which are thermometrically decoded, is depicted in Figure 16. As may be noticed, the switching sequence is anti-symmetrical in respect to the vertical axis of symmetry.

The ten most significant bits of the input bit string of the D/A converter of the invention are thermometrically decoded, while the four least significant bits select as many conducting branches of a binary scaled array.

CLAIMS

1. A method of generating a switching sequence for an unary array of conducting branches of a thermometrically decoded digital-to-analog converter, the conducting branches of which are affected by an error distributed over the array according to a certain error distribution function, comprising the steps of:

preliminary evaluating said error distribution function, fixing an upper bound function and a lower bound function symmetrical to each other of the INL error function associated to the switching sequence, determining the center of gravity of the array and calculating, by said preliminarily evaluated error distribution function, error values associated to each pair of conducting branches symmetrical in respect to the center of gravity of the array;

choosing pairs of successive conducting branches in the switching sequence such that every even conducting branch (2n) is substantially symmetrical to the preceding odd conducting branch (2n-1) in respect to said center of gravity;

building a switching sequence with an INL error function comprised between said upper bound and lower bound functions by:

- a) starting from a first chosen pair of conducting branches (1, 2) of the switching sequence iteratively performing the following steps from b1) to b3) for all other pairs of conducting branches of the array:
- b1) calculating a corresponding value of the INL error function of the switching sequence being built,
- b2) choosing as the successive pair, the pair of conducting branches that maximizes or minimizes the next value of the INL error function of the switching sequence though remaining comprised between the corresponding values of said upper bound and lower bound functions,
- b3) if all said other pairs do not meet the conditions of point b2), then repeating the steps from a) to b3) choosing every time a different first pair of branches, and if the conditions cannot yet be met, changing at

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- least one of said bound functions and restarting from point a);
- c) when all other pairs of conducting branches of the unary array meet the conditions, outputting the resulting switching sequence.
- 2. The method of claim 1, further comprising the operations of:

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- repeating the steps from a) to c) for all possible first pairs of conducting branches using the same bound functions, obtaining a set of switching sequences whose INL error function is comprised between said bound functions;
- choosing the switching sequence of said set that satisfies a pre-established criterion.
 - 3. The method of claim 1, wherein said bound functions are constant.
- 4. The method of claim 3, wherein the value of said upper bound function is half the maximum value of the absolute value of the DNL error affecting the conducting branches of the array.
- 5. The method of claim 1, wherein said bound functions are closer to 0 in correspondence of mid way values of said switching sequence than in correspondence of the two ends of a range of conversion.
 - 6. The method of claim 2, wherein said criterion consists in choosing the switching sequence of said set that is affected by the smallest absolute INL error.
- 7. The method of claim 2, wherein said criterion consists in choosing the switching sequence of said set that is affected by the smallest absolute DNL error in correspondence of mid way values of the switching sequence.
 - 8. The method of claim 2, wherein said criterion consists in choosing the switching sequence of said set that is affected by the most evenly oscillating INL error function.
 - 9. The method of claim 1, wherein said pairs of conducting branches are also chosen to make the switching sequence being built anti-symmetrical in

respect to an axis of symmetry of the unary array.

- 10. A thermometrically decoded digital/analog converter, comprising an unary array of conducting branches selectable by respective switches, characterized in that the switching sequence of said branches is generated by using the method of claim 1.
- 11. A computer program loadable in an internal memory of a computer, comprising a software code for performing the steps of the method of claim 1 when said program is executed on a computer.

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"METHOD OF GENERATING A SWITCHING SEQUENCE FOR AN UNARY ARRAY OF CONDUCTING BRANCHES AND A RELATIVE THERMOMETRICALLY DECODED DIGITAL-TO-ANALOG CONVERTER"

ABSTRACT

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A method for determining switching sequences of two-dimensional unary arrays of conducting branches of thermometrically decoded D/A converters, in a way that will ensure that the relative INL error function be contained between preestablished symmetrical upper and lower bound functions, has been found. When these upper and lower bound functions are constant, the obtained switching sequence compensates both the linear and the quadratic component of the error distribution and therefore is affected by a very small absolute INL error, which depends essentially on the random component of the error distribution.

This method may be easily implemented by a computer program and allows the realization of thermometrically decoded D/A converters affected by a known limited INL error function.

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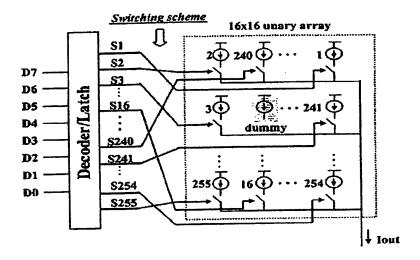


Fig. 1

Switch position	1	2	3	4	5	6	7	8
ε (%)	-7	-5	-3	-1	1	3_	5	7
Sequential sw. seq.	1	2	3	4	5	6	7	8
ε (%)	-7	-5	-3	-1	1	3	5	7
INL (%)	-7	-12	-15	<u>-16</u>	-15	-12	-7	0
Symmetrical sw. seq.	7	5	3	1	2	4	6	8
ε (%)	-1	+1	-3	+3	-5	+5	-7	+7
INL (%)	+1	0	+3	0	+5	0	+7	0

TAB. 1

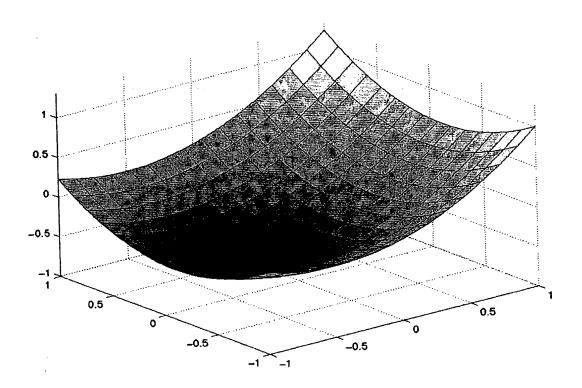
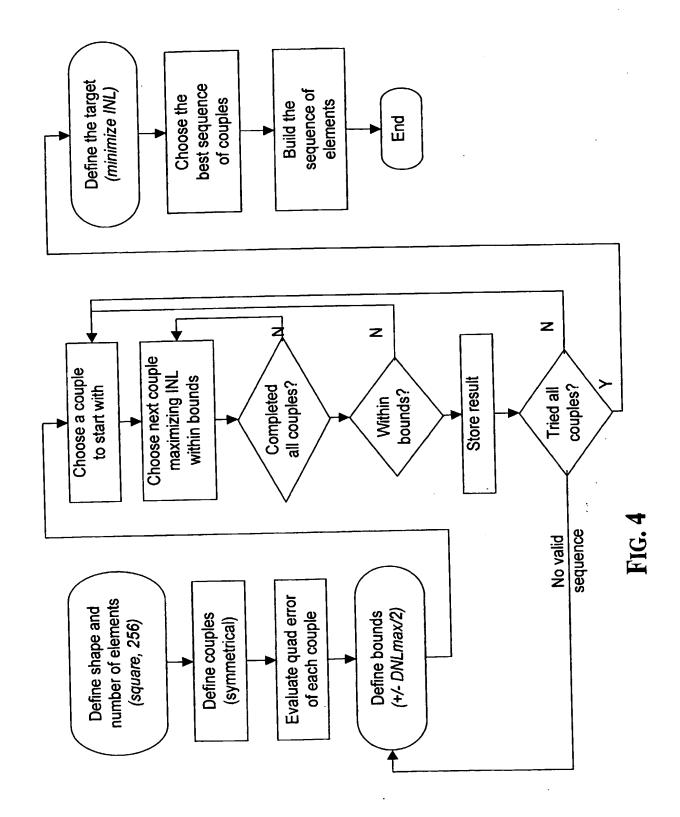


Fig. 2

a			
		g	
	h		
			b

FIG. 3



,	249	243	231	213	193	255	171	163	162	168	186	190	206	226	238	∞
	19	219	201	183	157	143	215	500	2	138	140	2 2	180	198	38	4
	3	29	175	149	131	233	251	155	152	228	222	130	148	82	82	56
	21	11	109	135	145	245	169	177	174	166	240	142	122	108	74	4
	29	103	127	115	159	187	195	207	204	192	182	86	112	126	9	9
	7	117	35	17	75	199	217	229	224	212	28	20	12	24	114	7
	68	119	58	91	65	45	235	247	242	73	6	62	88	9	45	88
	92	123	105	8	23	33	15	253	4	5	78	20	8	102	84	98
	93	47	101	6/	64	27	თ	က	254	91	8	22	8	106	124	96
	82	14	2	87	19	39	21	241	248	236	46	99	95	8	120	06
	-	113	23	±	69	27	211	223	230	218	200	92	18	98	118	72
	ន	66	125	111	97	181	191	203	208	196	188	160	116	128	\$	89
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FIG. 5

, 5	2	32	84	2	80	96	112	128	1	160	176	192	208	224	240	256
ቪ	2	31	47	83	62	92	£	127	143	159	175	191	207	223	239	255
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	9	22	38	22	2	88	102	118	2	150	99	182	198	214	230	246
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FIG. 6

	225	193	161	129	97	65	33	-	32	2	96	128	160	192	224	526
	227	195	163	131	66	29	35	က	30	62	95	126	158	190	222	254
	229	197	165	133	101	69	37	လ	78	8	95	124	156	188	220	252
	231	199	167	135	103	7	33	7	56	28	06	122	1 54	186	218	250
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	235	203	171	139	107	75	43	=	22	\$	98	118	150	182	214	246
	237	205	173	141	109	11	45	13	70	25	\$	116	148	180	212	244
	239	207	175	143	11	79	47	15	18	20	85	41	146	178	210	242
	241	500	171	145	113	₩	64	17	16	84	8	112	441	176	208	240
	243	211	179	147	115	83	51	19	4	4	78	110	142	174	206	238
;	245	213	181	149	117	82	53	72	12	4	92	108	140	172	204	236
	247	215	183	151	119	87	22	23	10	42	74	92	138	170	202	234
_	249	217	185	153	121	68	24	25	œ	9	22	\$	136	168	200	232
ın Symr	251	219	187	155	123	9	29	27	9	88	2	102	134	166	198	230
Seq Row-Column Symm	253	221	189	157	125	8	. 19	53	4	. 6	8 , 8	9	132	1 64	196	228
Seq Ro	255	223	191	159	127	55	8 8	; E	; ^	ן א	8	8 86	130	162	194	226

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	186	178	170	162	129	137	145	153	159	151	143	135	168	176	<u>\$</u>	192
	220	242	234	226	193	201	509	217	223	215	207	199	232	240	248	256
	254	246	238	230	197	205	213	22	219	211	203	195	228	236	244	252
	190	182	174	166	133	141	149	157	155	147	139	131	491	172	180	188
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သူ	158	150	142	<u>\$</u>	165	173	181	189	187	179	171	163	132	4	148	156
12-bit DAC	222	214	506	198	229	237	245	253	251	243	235	727	196	204	212	220

FIG. 8

223	191	127	8	210	178	+	20	214	182	#	5 2	218	8	12	28	
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47	#	175	207	¥	85	162	<u>\$</u>	38	102	166	198	45	106	170	202	
212	180	116	25	221	189	125	6	216	184	120	26	220	188	124	99	
148	70	244	2	157	59	253	93	152	54	248	88	156	78	252	95	
89	228	4	132	11	237	13	141	72	232	œ	136	92	236	12	140	
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155	27	251	9	153	25	249	88	145	17	241	8	158	စ္က	254	8	
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43	107	171	203	4	105	169	201	33	26	161	193	94	110	174	206	
215	183	119	55	211	179	115	51	224	192	128	2	213	181	117	જુ	
151	23	247	87	147	19	243	8	160	32	256	96	149	21	245	82	
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Q2 Random Walk

NL-bounded	papu												Ş	6	Ç
206	174	9	62	195	163	19	21	199	167	23	22	196	164	3	70
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82	4	158	190	29	35	147	179	71	39	151	183	88	೫	148	180
193	161	17	49	208	176	32	2	205	173	59	19	204	172	78	09
225	129	113	8	240	<u>‡</u>	128	96	237	141	125	93	236	140	124	35
26		241	209	112	16	256	224	109	5	253	221	108	12	252	220
65	33	145	171	80	84	160	192	11	. 45	157	189	9/	4	156	188
201	169	. 52	22	197	165	51	ន	202	170	56	88	200	168	24	99
233	137	121	89	229	133	117	82	234	138	122	06	232	136	120	88
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73	4	153	185	69	37	149	181	47	42	154	186	72	40	152	1 84
203	171	27	29	198	166	22	ጃ	19	162	8	20	207	175	સ	63
235	139	123	9	230	134	118	88	226	130	114	83	239	143	127	92
107	=	251	219	102	ဖ	246	214	88	7	242	210	1	15	255	223
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FIG. 10

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213	147	101	35	210	18	254	95	11	က	224	134	145	75	116	231
22	506	135	3	162	36	149	62	35	7	107	189	251	168	105	178
219	19	∞	83	194	27	170	146	14	124	125	152	53	ጷ	237	26
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244	39	26	247	202	\$	236	121	122	23	42	217	79	29	c,	38
28	212	73	2	157	129	203	150	79	139	4	142	166	230	84	195
191	126	132	97	140	901	104	45	248	196	श्र	218	240	153	89	29
220	47	51	205	185	5	G	7	117	128	211	8	8	12	199	155
118	169	246	173	225	207	136	180	8	232	222	82	253	28	12	160
123	209	22	8	114	9	242	214	20	233	2	216	4	192	65	252
130	182	143	154	9	119	198	74	159	201	70	7	49	127	188	20
151	103	25	172	239	06	33	2 8	243	255	87	32	15	37	112	109

Magic														,	;
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33	223	222	36	37	219	218	04	4	215	214	4	45	211	210	84
208	20	51	205	50	%	55	201	700	88	29	197	96	62	83	193
192	99	29	189	188	02	71	185	184	74	75	181	180	82	79	171
84	175	174	84	82	171	170	88	68	167	166	95	83	163	162	8
26	159	158	6	101	155	154	₹	105	151	150	108	109	147	146	112
4	<u>4</u>	115	14	140	118	119	137	136	122	123	133	132	126	127	129
128	130	131	125	124	<u>\$</u>	135	121	120	138	139	117	116	142	143	113
145	Ξ	110	148	149	107	106	152	53	103	102	156	157	66	86	160
161	92	8	164	165	91	06	168	169	87	98	172	173	83	82	176
80	178	179	11	92	182	183	73	72	186	187	69	89	190	191	92
2	2	195	61	09	198	199	22	26	202	203	ಜ	25	506	207	49
509	47	46	212	213	43	42	216	217	88	88	220	221	35	क्ष	224
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Modified Magic 253 9	237	25	33	213	64	197	254	6	238	92	8	214	20	198
8		165	157	105	141	121	99	182	83	166	158	901	142	122
109		153	161	82	177	69	126	138	110	<u>\$</u>	162	98	178	29
209		37	59	233	5	249	194	22	210	88	30	234	4	250
21		225	217	45	201	19	9	242	23	226	218	94	202	62
169		93	101	145	117	129	186	78	170	\$	102	146	118	130
149		26	88	173	73	189	1 34	114	120	88	8	174	4	190
4		22	229	17	245	-	28	206	45	222	230	8	246	7
239		27	35	215	21	199	256	12	240	78	မွ	216	25	200
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11		155	163	87	179	71	128	140	112	156	2	88	180	72
211		39	34	235	15	251	196	26	212	4	32	236	9	252
23		227	219	47	203	83	∞	244	54	228	220	84	204	2
171		92	103	147	119	131	188	80	172	8	\$	148	120	132
151		66	9	175	75	191	136	116	152	9	95	176	92	192
4		223	231	19	247	က	99	208	4	224	232	70	248	4

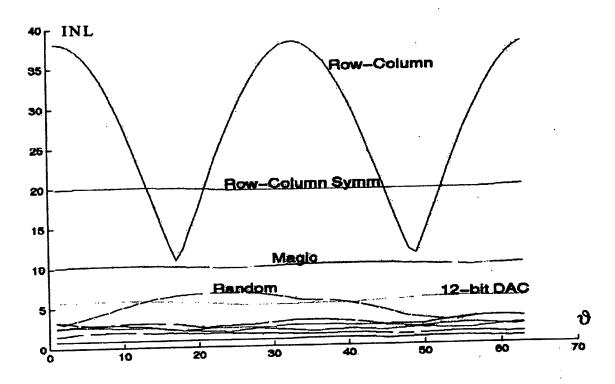


FIG. 14

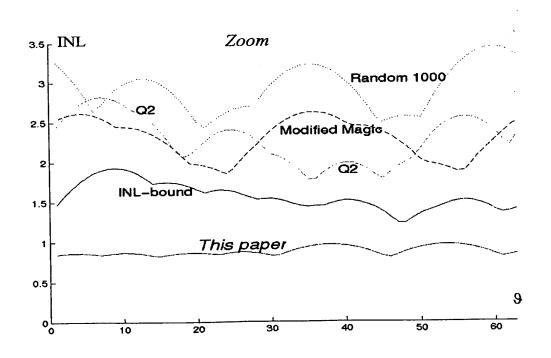


FIG. 15

The same of the sa	16 x 16 matrix comparison result	$(g = g_q = 0.5)$
	Algorithm/Sequence	Maximum absolute INL
Row-Col	Row-Column basic	38.1
KOW-COI	Row-Column symm. Seq.	20
	Row-Column hier symm seq. [2]	10.2
Hierarchic.	12-bit C05 DAC	5.9
	Q2 random walk [1]	2.8
	Hier. switch, INL-bounded [2]	1.93
Improvements	Random (1 trial)	6.8
Improvements	Random (best of 1k trials)	3.4
•	Magic	10.2
	Modified Magic	2.6
	This work	0.96

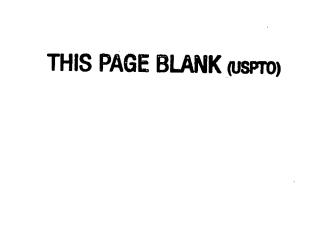
TAB. 2

Maximum absolute INL error			
$g_i =$	0.3	0.5	0.7
$g_q = \frac{1}{2}$	0.7	0.5	0.3
Q ² rand walk	2.26	2.8	3.38
INL-bounded	1.56	1.93	2.4
This work	0.95	0.96	1.04

TAB. 3

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Frc. 16



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